Astronomical Techniques II Lecture 14 - Sensitivity and a few misc. topics

Divya Oberoi

IUCAA NCRA Graduate School

div@ncra.tifr.res.in

March-May 2016

▲ロト ▲圖ト ▲ヨト ▲ヨト ニヨー のへで

1/21

Noise and Temperature

Sensitivity of a 2 element interferometer

$$\begin{array}{l} 1 < P_i >= a_i < (s_i + n_i)^2 >= a_i [< s_i >^2 + < n_i >^2] \\ 2 < P_i >= g_i^2 k_B (T_{a \ i} + T_{sys \ i}) \Delta \nu \\ 3 < P_i >= g_i^2 k_B (K_i \ S_T + T_{sys \ i}) \Delta \nu \\ 4 < P_{ij} >= \frac{g_i g_j}{\eta_s} \sqrt{K_i K_j} k_B \Delta \nu S_c \end{array}$$

3/21

Sensitivity of a 2 element interferometer

SNR - ratio of DC component to the RMS fluctuations of the correlator output

$$\sum \Delta S_{ij} = \frac{1}{\eta_s \sqrt{2 \Delta \nu \tau_{acc}}} \sqrt{S_c^2 + S_T^2 + S_T (\frac{T_{sys i}}{K_i} + \frac{T_{sys j}}{K_j}) + \frac{T_{sys i} T_{sys j}}{K_i K_j}}$$

Assumes a square bandpass, but can be generalized to an arbitrary bandpass

Sensitivity of a 2 element interferometer

1 Weak source case
$$S_T << \frac{T_{sys}}{K}$$

$$\Delta S_{ij} = \frac{1}{\eta_s} \sqrt{\frac{T_{sys \ i} T_{sys \ j}}{2 \ \Delta \nu \ \tau_{acc} \ K_i K_j}} = \frac{1}{\eta_s} \sqrt{\frac{SEFD_i \ SEFD_j}{2 \ \Delta \nu \ \tau_{acc}}}$$
2 Strong source case $S_T >> \frac{T_{sys}}{K}$

$$\Delta S_{ij} = \frac{S_T}{\eta_s \sqrt{2 \ \Delta \nu \ \tau_{acc}}}$$
1 Usually $S_T >> S_C$

< □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □ > < □

Amplitudes and Phases

$$S_m = \sqrt{S_R^2 + S_I^2}$$

$$\phi_m = tan^{-1} \frac{S_I}{S_R}$$
Noise distribution for S_m - Rice distribution
$$P(S_m) = \frac{S_m}{\Delta S^2} l_0 \left(\frac{S_m S}{\Delta S^2}\right) e^{\frac{-(S_m^2 + S^2)}{2\Delta S^2}}$$
where l_0 is the modified Bessels function of the first kind, order zero, and S is the true amplitude.
Probability distribution for phase error $\phi - \phi_m$, where ϕ is the true phase
$$P(\phi - \phi_m) = \frac{1}{2\pi} e^{\frac{-S^2}{2\Delta S^2}} \left(1 + G\sqrt{\pi}e^{G^2}(1 + erfG)\right)$$

where
$$G(\theta) = \frac{Scos\theta}{\sqrt{2}\Delta S}$$

Probability distribution of measured amplitude and phase



つくで 7/21

Sensitivity for a point source

$$\Delta I_m = \frac{\sqrt{2}k_B T_{sys}}{\eta_c \eta_a A \sqrt{N_{base} \Delta \nu N_{times}}}$$

$$\eta_c = \frac{Sensitivity \ of \ the \ correlator}{Sensitivity \ of \ a \ perfect \ analog \ correlator}$$

$$1 \ bit - 64\%; \ 2 \ bit \ 3 \ level - 81\%$$

$$\Delta I_m = \frac{1}{\eta_c} \frac{SEFD}{\sqrt{N(N-1) \ \Delta \nu \ N_{times}}}$$

Effect of the primary beam

1
$$I_m(l,m) = I(l,m) P(l,m) + N(l,m)$$

2 $\frac{I_m(l,m)}{P(l,m)} = I(l,m) + \frac{N(l,m)}{P(l,m)}$

A formalism for 3-D imaging

 $V(u, v, w) = \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} \frac{\mathcal{A}(l, m) \, l(l, m)}{\sqrt{1 - l^2 - m^2}} e^{-2\pi i (ul + vm + w(\sqrt{1 - l^2 - m^2} - 1))} \, dl \, dm$

2

$$V(u, v, w) e^{-2\pi i w} = \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} \frac{\mathcal{A}(l, m) l(l, m)}{\sqrt{1 - l^2 - m^2}}$$
$$\delta(n - \sqrt{1 - l^2 - m^2})$$
$$e^{-2\pi i (ul + vm + wn)} dl_0 dm dn \in \mathbb{R} \xrightarrow{\infty} \frac{10}{10/2}$$

$$I^{D(3)} = \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} V(u, v, w) S(u, v, w) e^{-2\pi i w}$$
$$e^{2\pi i (ul + vm + wn)} du dv dw$$

 $I^{D(3)} = I^{(3)} \star B^{D(3)}$ where, $I^{(3)}(I, m, n) = \frac{\mathcal{A}(I, m) I(I, m)}{\sqrt{1 - l^2 - m^2}} \delta(n - \sqrt{1 - l^2 - m^2})$

3D Imaging



Figure 19-1. The image volume and its relation to the sky brightness. (Left) Threedimensional transformation of the analytic visibility function maps the sky brightness onto a unit sphere. The dots represent these sources. (Middle) Convolution with a dirty beam results in sidelobes, shown as dashed lines, throughout the volume above and below the unit sphere. (Right) After deconvolution, the images are represented by finite-size "clean beams" on the unit sphere. The two-dimensional image is recovered by projection onto the tangent plane, indicated by vertical dashed lines.

3D Imaging



Figure 19-2. The image volume and its relation to a 'standard' two-dimensional image. (Left) At a particular time, a 'snapshot' with a two-dimensional array will project the true structure on the unit sphere onto the tangent plane with a 'ray beam', tilted at a particular angle given by the geometry of the array at the time of observation. (Right) At a later time, the array geometry has changed due to earth rotation, so the projection is now at a different angle. The apparent positions of the objects which are not located at the tangent point have changed with respect to the earlier observation.

3D Imaging

1 Faceting/Polyhedron imaging

- Divide the image into many many facets, each small enough that the small FoV and small w term approximation are sastified within it
- 2 CLEAN flux is subtracted from ungridded visibilities
- 3 No. of facets depends upon FoV and resolution
- **4** 100-1000 times slower than 2D imaging
- 2 w projection (Cornwell, Golap and Bhatnagar, 2008)

1 V(u, v, w) =
$$\int_{-\infty}^{\infty} \int_{-\infty}^{\infty} \frac{l(l, m)}{\sqrt{1 - l^2 - m^2}} \mathbf{G}(l, m, w) e^{-2\pi i(ul + vm)} \, dl \, dm,$$
where
$$\mathbf{G}(l, m, w) = e^{-2\pi i(w\sqrt{1 - l^2 - m^2} - 1)}$$

2 V(u, v, w) = $\hat{G}(u, v, w) \bigstar V(u, v, w = 0)$
3 Order of magnitude faster

Polarization

- Most non-thermal processes give rise to at least partially polarised emission
- Polarized emission is an important diagnostic of the conditions in the radio source and in the intervening medium

- Additional DoF needed to describe the polarization state of radiation
- A given probe is sensitive to only one of the orthogonal pols (linear or circular)
- 6 Measure both polarizations and compute all four cross-correlations

Polarization Measurements

1 Significantly harder than total intensity

- For the vast majority of sources, fractional polarization is quite low - pushed into low SNR regimes
- 2 Number of DoF for imaging increase by a factor of 4
- 3 Calibration issues
 - 1 Instrumental
 - 2 Propagation
 - **3** Need for polarization calibrator
 - Calibration tends to have a strong direction dependence (absolute, as well as within the fov)

▲□▶ ▲□▶ ▲三▶ ▲三▶ 三三 ろのの

16/21

5 Alt-Az mounts

The Hamaker-Bregman-Sault Measurement Equation

- 1 Hamaker, Bregman and Sault 1996-1998
- 2 Jones Matrix

$$E_{0} \cos(\omega t + \phi) = E_{0}e^{i\phi}$$

$$\begin{bmatrix} E_{R} \\ E_{L}' \end{bmatrix} = \begin{pmatrix} a & b \\ c & d \end{pmatrix} \begin{pmatrix} E_{R} \\ E_{L} \end{bmatrix}$$

$$\begin{bmatrix} J_{gain} = \begin{pmatrix} g_{R} & 0 \\ 0 & g_{L} \end{pmatrix}$$

$$\begin{bmatrix} J_{leakage} = \begin{pmatrix} 1 & D_{R} \\ -D_{L} & 1 \end{pmatrix}$$

$$\begin{bmatrix} J_{rotation} = \begin{pmatrix} e^{-i\theta} & 0 \\ 0 & e^{i\theta} \end{pmatrix}$$

$$\begin{bmatrix} J_{overall} = J_{gain} & J_{leakage} & J_{rotation} \dots$$

Jones matrices

- Js are different for each antenna, and are usually time and frequency dependent
- Provide a framework to represent propagation of signal path up to the correlator
- 3 Complicated systems can be handled gracefully
- Provides an approach which allows individual effects to be modelled in different physically relevant manners
- Matrix formulation well suited for computational scalability and efficiency

Jone matrices - Polarimetric Equivalent

1
$$V'_{i,j} = g_i g_j^* V_{i,j}$$

2 $\mathbf{A} \bigotimes \mathbf{B} = a_{i,j} \mathbf{B}$

$$(\mathbf{A}_i\mathbf{B}_i)\otimes(\mathbf{A}_j\mathbf{B}_j)=(\mathbf{A}_i\otimes\mathbf{A}_j)(\mathbf{B}_i\otimes\mathbf{B}_j)$$

3 Inputs to the correlator -
$$E'_i = \mathbf{J}_i E_i$$

4 Outputs of the correlator - $E'_i \bigotimes E'^*_j$

$$(\mathbf{J}_{i}E_{i}) \otimes (\mathbf{J}_{j}E_{j})^{*} = (\mathbf{J}_{i} \otimes \mathbf{J}_{j}^{*})(E_{i} \otimes E_{j}^{*})$$

$$\mathbf{E}_{k,i}E_{k,j}^{*} = \begin{pmatrix} E_{R,i}E_{R,j}^{*} \\ E_{R,i}E_{L,j}^{*} \\ E_{L,i}E_{R,j}^{*} \\ E_{L,i}E_{L,j}^{*} \end{pmatrix}$$

$$\mathbf{I} < E'_{i} \bigotimes E'^{*}_{j} >= \begin{pmatrix} V_{RR,ij} \\ V_{RL,ij} \\ V_{LR,ij} \\ V_{LL,ij} \end{pmatrix}$$

2 $V'_{ij} = (\mathbf{J}_i \bigotimes \mathbf{J}_j^*) V_{ij}$ - V_{ij} - Coherency vector

 Calibration requires estimating the different J_is and applying the inverse matrix to the measured Coherency vector

Relationship with Stokes vectors

$$\mathbf{I} \quad V_{S} = \begin{pmatrix} V_{I} \\ V_{Q} \\ V_{U} \\ V_{V} \end{pmatrix} - \text{Stokes Visibility Vector}$$

$$\mathbf{I} \quad V_{ij} = \mathbf{S} \quad V_{S,ij}$$

$$\mathbf{I} \quad V'_{ij} = (\mathbf{J}_{i} \bigotimes \mathbf{J}_{j}) \quad \mathbf{S} \quad V_{S,ij}$$

$$\mathbf{I} \quad \mathbf{S}_{circ} = \begin{pmatrix} 1 & 0 & 0 & 1 \\ 0 & 1 & i & 0 \\ 0 & 1 & -i & 0 \\ 1 & 0 & 0 & -1 \end{pmatrix} \quad \mathbf{S}_{linear} = \begin{pmatrix} 1 & 1 & 0 & 0 \\ 0 & 0 & 1 & i \\ 0 & 0 & 1 & -i \\ 1 & -1 & 0 & 0 \end{pmatrix}$$